## 10.7

*Solution.* The poles of the *z*-transform are

$$z_1 = \frac{1}{2}j, \qquad z_2 = -\frac{1}{2}j, \qquad z_3 = \frac{3}{4}, \qquad z_4 = -\frac{1}{2}$$

Based on these poles, the ROC can be

$$0 < \|z\| < \frac{1}{2}, \qquad \frac{1}{2} < \|z\| < \frac{3}{4}, \qquad {\rm or} \ \frac{3}{4} < \|z\|.$$

## 10.16

Note that, for a stable discrete-time LTI system with rational system function H(z), the condition (a) for causality is equivalent to "the outermost pole is inside the unit circle".

(a)

Solution. The given *z*-transform has a pole at infinity. Therefore, it is not causal.

#### **(b)**

**Solution.** The poles of the z-transforklkjm are at  $z = \frac{1}{4}$  and  $z = -\frac{3}{4}$ . Therefore, it is causal.

#### (c)

**Solution.** The given z-transform has a pole at  $z = \frac{4}{3}$ . Therefore, it is not causal.

## 10.21

### (a)

**Solution.** For  $x[n] = \delta[n+5]$ ,

$$X(z) = z^5, \qquad \text{All } z,$$

as shown in the figure below.

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TODO
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## (g)

**Solution.** For  $x[n] = 2^n u[-n] + \frac{1}{4^n} u[n-1]$ , by table 10.1 and 10.2,

$$X(z) = \frac{2}{2-z} + \frac{1}{4z-1}, \qquad \frac{1}{4} < \|z\| < 2,$$

as shown in the figure below.

TODO

# 10.24

# (a)

Solution. We have

$$X(z) = \frac{1 - 2z^{-1}}{\left(1 + \frac{1}{2}z^{-1}\right)\left(1 + 2z^{-1}\right)} = \frac{8}{3(1 + 2z^{-1})} - \frac{5}{3\left(1 + \frac{1}{2}z^{-1}\right)}.$$

Since x[n] is absolutely summable, the ROC is  $\frac{1}{2} < \|z\| < 2.$  It follows that

$$x[n] = \frac{8}{3} \left(-\frac{1}{2}\right)^n u[n] - \frac{5}{3} (-2)^n u[n].$$

**(b)** 

Solution. We have

$$X(z) = 1 - z^{-1} + \frac{1}{2}z^{-2} - \frac{1}{4}z^{-3} + \dots = 1 - \sum_{n=1}^{\infty} \left(-\frac{1}{2}\right)^{n-1} z^{-n}.$$

Using the analysis equation, we get

$$x[n] = \delta[n] - \left(-\frac{1}{2}\right)^{n-1} u[n-1]$$

## 10.33

(a)

Solution. Taking the *z*-transform of the equation and simplifying, we get

$$H(z) = \frac{Y(z)}{X(z)} = \frac{1}{1 - \frac{1}{2}z^{-1} + \frac{1}{4}z^{-2}}.$$

## **(b)**

**Solution.** If  $x[n] = \left(\frac{1}{2}\right)^n u[n]$ , then

$$X(z) = \frac{1}{1 - \frac{1}{2}z^{-1}}, \qquad \|z\| > \frac{1}{2}.$$

Hence

$$Y(z) = H(z)X(z) = \frac{1}{1 - \frac{1}{2}z^{-1}} + \frac{\frac{1}{2}z^{-1}}{1 - \frac{1}{2}z^{-1} + \frac{1}{4}z^{-2}}.$$

Using table 10.2, we get

$$y[n] = \left(\frac{1}{2}\right)^n u[n] + \frac{2}{\sqrt{3}} \left(\frac{1}{2}\right)^n \sin\left(\frac{\pi}{3}n\right) u[n].$$

## 10.37

## (a)

**Solution.** Let w[n] denote the signal represented by the top-middle node. Then from the diagram we know that

$$\begin{split} x[n] &= w[n] + \frac{1}{3}w[n-1] - \frac{2}{9}w[n-2], \\ y[n] &= w[n] - \frac{9}{8}w[n-1]. \end{split}$$

Hence

$$H(z) = \frac{Y(z)}{X(z)} = \frac{\frac{Y(z)}{W(z)}}{\frac{X(z)}{W(z)}} = \frac{1 - \frac{9}{8}z^{-1}}{1 + \frac{1}{3}z^{-1} - \frac{2}{9}z^{-2}}.$$

Taking the inverse *z*-transform, we obtain

$$y[n] + \frac{1}{3}y[n-1] - \frac{2}{9}y[n-2] = x[n] - \frac{9}{8}x[n-1].$$

### **(b)**

**Solution.** The poles of the *z*-transform are at  $z = \frac{1}{3}$  and  $z = -\frac{2}{3}$ . Therefore, the system is stable.

### 10.42

### **(b)**

Solution. Applying the unilateral *z*-transform, we obtain

$$\mathcal{Y}(z) - \frac{1}{2}z^{-1}\mathcal{Y}(z) - \frac{1}{2}y[-1] = \mathcal{X}(z) - \frac{1}{2}z^{-1}\mathcal{X}(z).$$

For the zero-input response, setting  $\mathcal{X}(z) = 0$ , we get  $\mathcal{Y}(z) = 0$ . Then by applying the inverse unilateral z-transform, we obtain  $y_{zi}[n] = 0$ .

For the zero-state response, setting y[-1] = 0, we get  $\mathcal{Y}(z) = \mathcal{X}(z) = \frac{1}{1-z^{-1}}$ . Then by applying the inverse unilateral z-transform, we obtain  $y_{zs}[n] = u[n]$ .

### 10.59

#### (a)

**Solution.** Let w[n] denote the signal represented by the top-middle node. Then from the diagram we know that

$$\begin{split} x[n] &= w[n] + \frac{k}{3}w[n-1], \\ y[n] &= w[n] - \frac{k}{4}w[n-1]. \end{split}$$

Hence

$$H(z) = \frac{Y(z)}{X(z)} = \frac{\frac{Y(z)}{W(z)}}{\frac{X(z)}{W(z)}} = \frac{1 - \frac{k}{4}z^{-1}}{1 + \frac{k}{3}z^{-1}}.$$

Since H(z) correspond to a causal filter, the ROC is  $\frac{\|k\|}{3} < \|z\|$  , as shown in the figure below. TODO

# (b)

**Solution.** For the system to be stable, there must be  $\frac{\|k\|}{3} < 1$ , that is,  $\|k\| < 3$ .

## (c)

**Solution.** Since  $x[n] = \left(\frac{2}{3}\right)^n$ , we have

$$y[n] = H\left(\frac{2}{3}\right)x[n] = \frac{1 - \frac{1}{4}\left(\frac{2}{3}\right)^{-1}}{1 + \frac{1}{3}\left(\frac{2}{3}\right)^{-1}}\left(\frac{2}{3}\right)^n = \frac{5}{12}\left(\frac{2}{3}\right)^n.$$